

# Speech and Audio Coding Theory

## Contents of lecture

- ❑ Pitch estimation
  - ❑ Time domain methods
  - ❑ Frequency domain methods
  - ❑ Time- and frequency-domain methods
  - ❑ Pre- and post-processing techniques
- ❑ Voiced-unvoiced classification
  - ❑ Hard-decision voicing
  - ❑ Soft-decision voicing

# ***Introduction***

## ❑ Three main speech features

- ❑ Spectral envelope: from short-term correlation → LSFs

- ❑ Pitch (period and gains): from long-term correlation

  - ❑ Especially for pitch period

    - ❑ Used in pitch predictor, to reduce the search space for LTP parameters (gains)

    - ❑ Used in the generation of excitation signal for a voiced region

- ❑ Voiced-unvoiced (V-UV) classification

  - ❑ Voiced: high energy, periodicity

    - ❑ If incorrectly classified as unvoiced, the synthesized speech will sound rough and less intelligible.

  - ❑ Unvoiced: like random noise with no periodicity

    - ❑ If incorrectly classified as voiced, the synthesized speech will sound annoyingly metallic or robotic.

  - ❑ Transition region between voiced and unvoiced, or inherently mixed (i.e., /d/)

    - ❑ A soft decision voicing: frequency-band-dependent V-UV classification

    - ❑ The soft decision is usually carried out in the frequency domain.

# ***Pitch estimation***

- ❑ Why accurate and reliable pitch period estimation is difficult?
  - ❑ No perfect train of periodic pulses, even in voiced regions
    - ❑ slowly evolves from cycle to cycle
  - ❑ Onset and offset regions of voiced speech are not stationary.
  - ❑ In some parts, the speech may contain a mixture of voiced and unvoiced signals.
  - ❑ Interaction with 1<sup>st</sup> formant as in the child or female speech
  - ❑ Background ambient noise
- ❑ Pitch determination algorithms (PDA) based on
  - ❑ Time domain properties
  - ❑ Frequency domain properties
  - ❑ Both the time and frequency domain properties

## ***Time domain methods for PD***

- ❑ Idea: using similarity of the waveform in time domain
- ❑ AMDF (Average Magnitude Difference Function) PDA

❑ Definition: 
$$A(\tau) = \frac{1}{N} \sum_{n=0}^{N-1} |s(n) - s(n + \tau)|$$

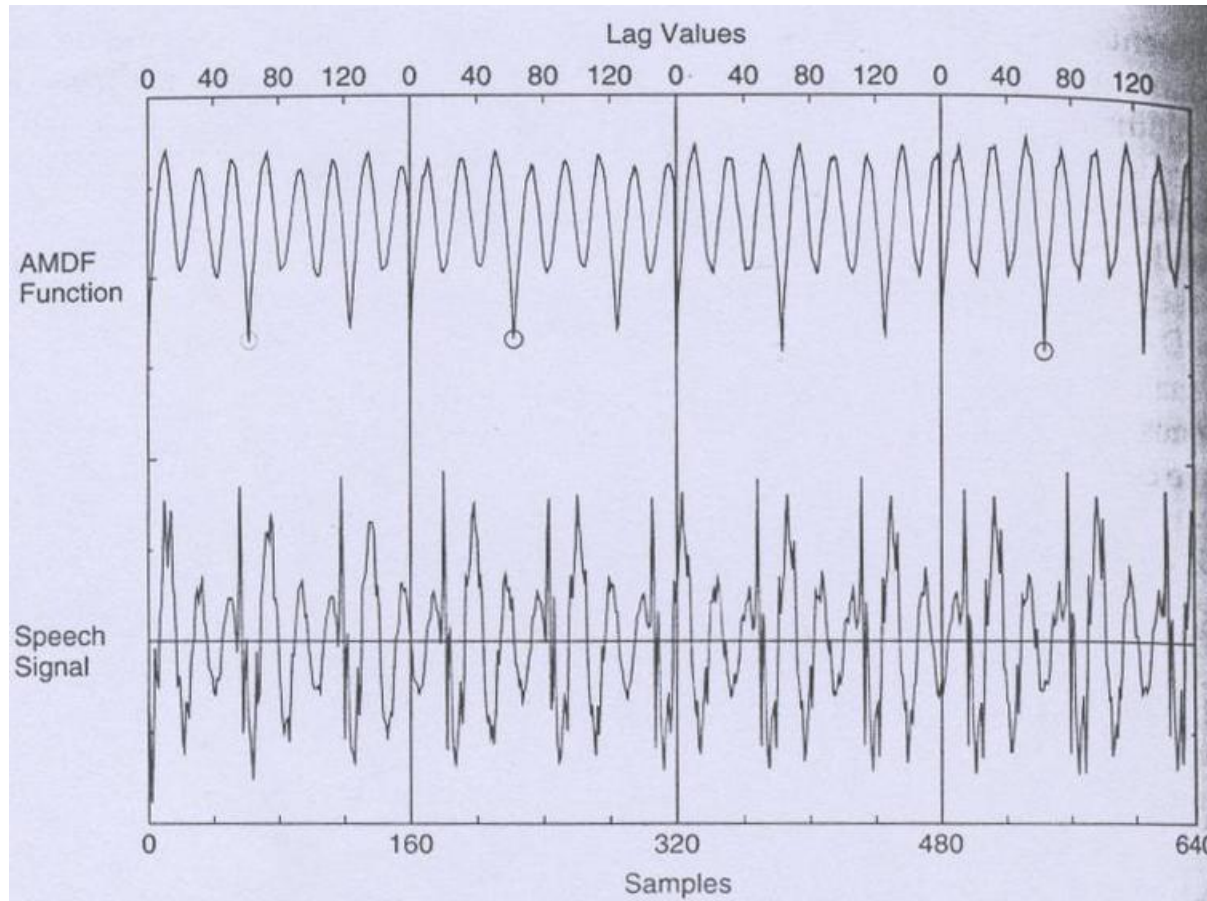
- ❑ Anti-correlation measure (dissimilarity measure)

### ❑ Merits

- ❑ Simple computation
- ❑ Not useful with DSPs optimized for multiplications and additions, but still useful with ASICs having no arithmetic component.
- ❑ Smaller dynamic range due to no multiplications
  - ❑ Bounded to zero
- ❑ Narrower valleys for stationary signals

## ***Time domain methods for PD***

- ❑ AMDF (Average Magnitude Difference Function) PDA



# ***Time domain methods for PD***

## ❑ Auto-correlation PDA

### ❑ Definition

$$❑ E(\tau) = \frac{1}{N} \sum_{n=0}^{N-1} [s(n) - s(n + \tau)]^2$$

### ❑ Normalized criterion reflecting the non-stationary effect of pitch

$$❑ E(\tau) = \frac{1}{N} \sum_{n=0}^{N-1} [s(n) - \beta s(n + \tau)]^2$$

❑  $\beta$ : scaling factor (pitch gain)

# ***Time domain methods for PD***

## **❑ Auto-correlation PDA**

❑ If the signal is stationary (that is,  $\sum s^2(n) = \sum s^2(n + \tau)$ ), the similarity function becomes  $E(\tau) = R(0) - R(\tau)$

❑ Here, 
$$R(\tau) = \sum_{n=0}^{N-1} s(n)s(n + \tau)$$

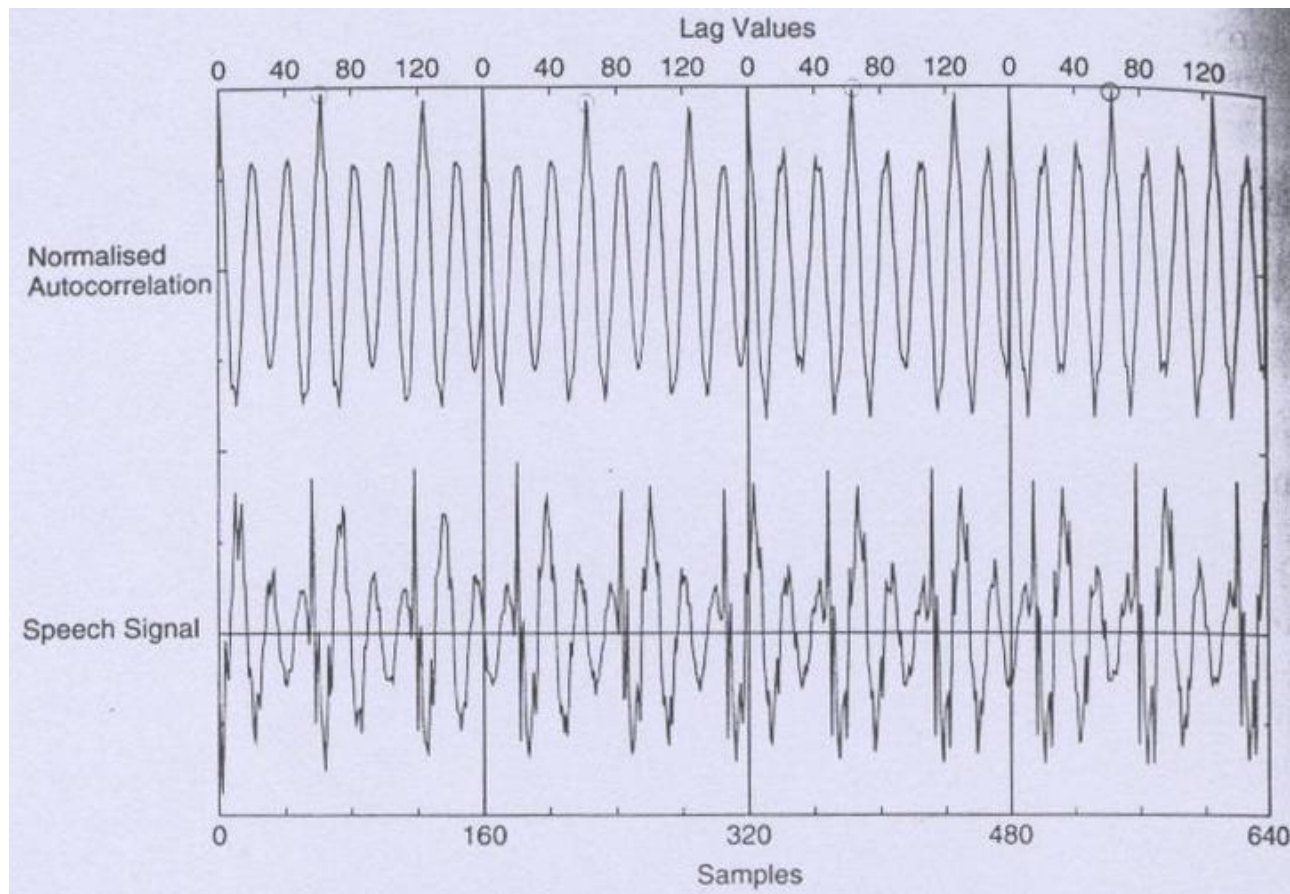
❑ Therefore, minimizing  $E(\tau)$  corresponds approximately to maximizing  $R(\tau)$ .

❑ Merits

- ❑ Easy to implement in real-time with DSPs due to its regular form of multiplications
- ❑ Phase insensitive

## ***Time domain methods for PD***

### **□ Auto-correlation PDA (Normalized version)**





# ***Time domain methods for PD***

## **❑ Auto-correlation PDA**

❑ Generalized similarity measure

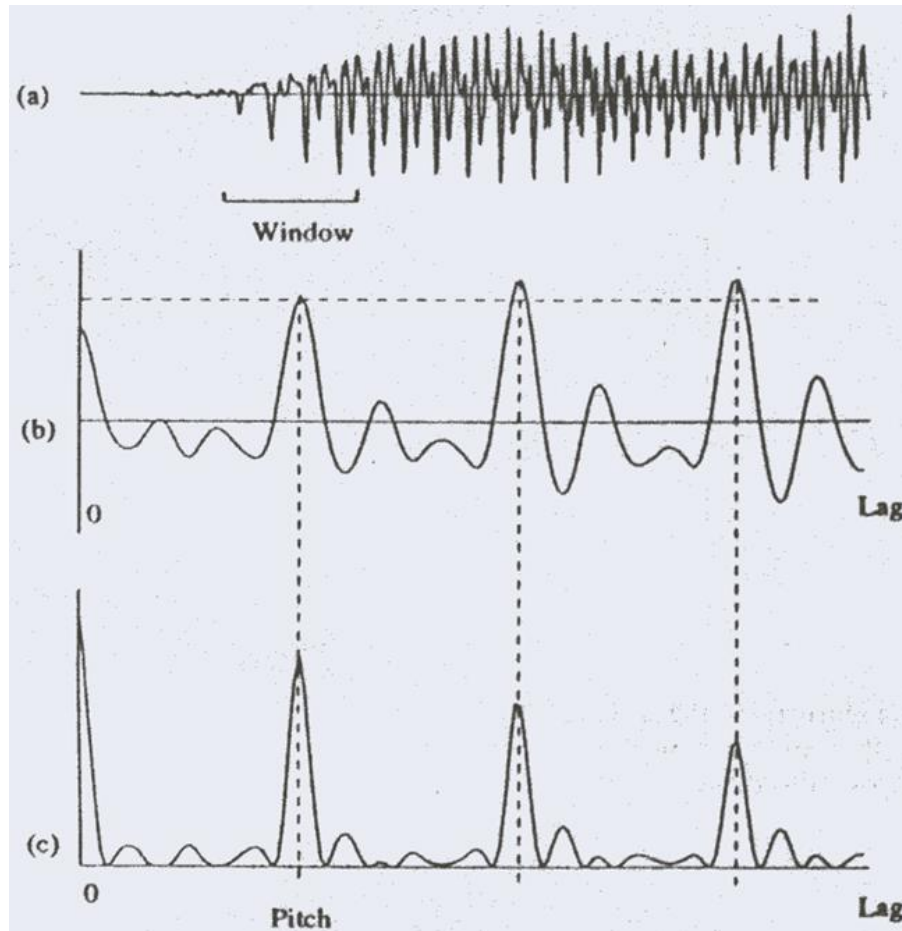
$$\text{❑ } E(\tau) = \frac{1}{N} \left\{ \sum_{n=0}^{N-1} |s(n) - s(n + \tau)|^k \right\}^{\frac{1}{k}}$$

❑ From experiments,  $k=2$  is best.

❑ Since it corresponds to the auto-correlation method, it means that auto-correlation method is superior to AMDF method.

# ***Time domain methods for PD***

## ❑ Drawback of the direct auto-correlation method



### (a) Original speech

- Window taken on an onset region

### (b) Direct autocorrelation function

- Difficult to set an appropriate TH

### (c) Normalized autocorrelation function

- Always a consistent pattern
  - Decreasing peaks
  - Bounded to zero
- Relatively easy to set the TH

# ***Time domain methods for PD***

## ❑ Normalized auto-correlation method

❑ From  $E(\tau) = \frac{1}{N} \sum_{n=0}^{N-1} [s(n) - \beta s(n + \tau)]^2$  and  $\partial E(\tau) / \partial \beta = 0$ ,

we get

$$\beta = \frac{\sum_{n=0}^{N-1} s(n)s(n + \tau)}{\sum_{n=0}^{N-1} s^2(n + \tau)}$$

❑ Substituting this to the normalized criterion, we obtain

$$E(\tau) = \sum_{n=0}^{N-1} s^2(n) - \frac{\left[ \sum_{n=0}^{N-1} s(n)s(n + \tau) \right]^2}{\sum_{n=0}^{N-1} s^2(n + \tau)}$$

❑ Removing the negative correlation effects, the criterion becomes to maximize

$$R(\tau) = \frac{\sum_{n=0}^{N-1} s(n)s(n + \tau)}{\sqrt{\sum_{n=0}^{N-1} s^2(n + \tau)}}$$

# ***Frequency domain methods for PD***

## **❑ Basic idea**

- ❑ Using the harmonic structure in frequency domain
- ❑ Main drawback: high computational complexity

## **❑ Harmonic peak detection method**

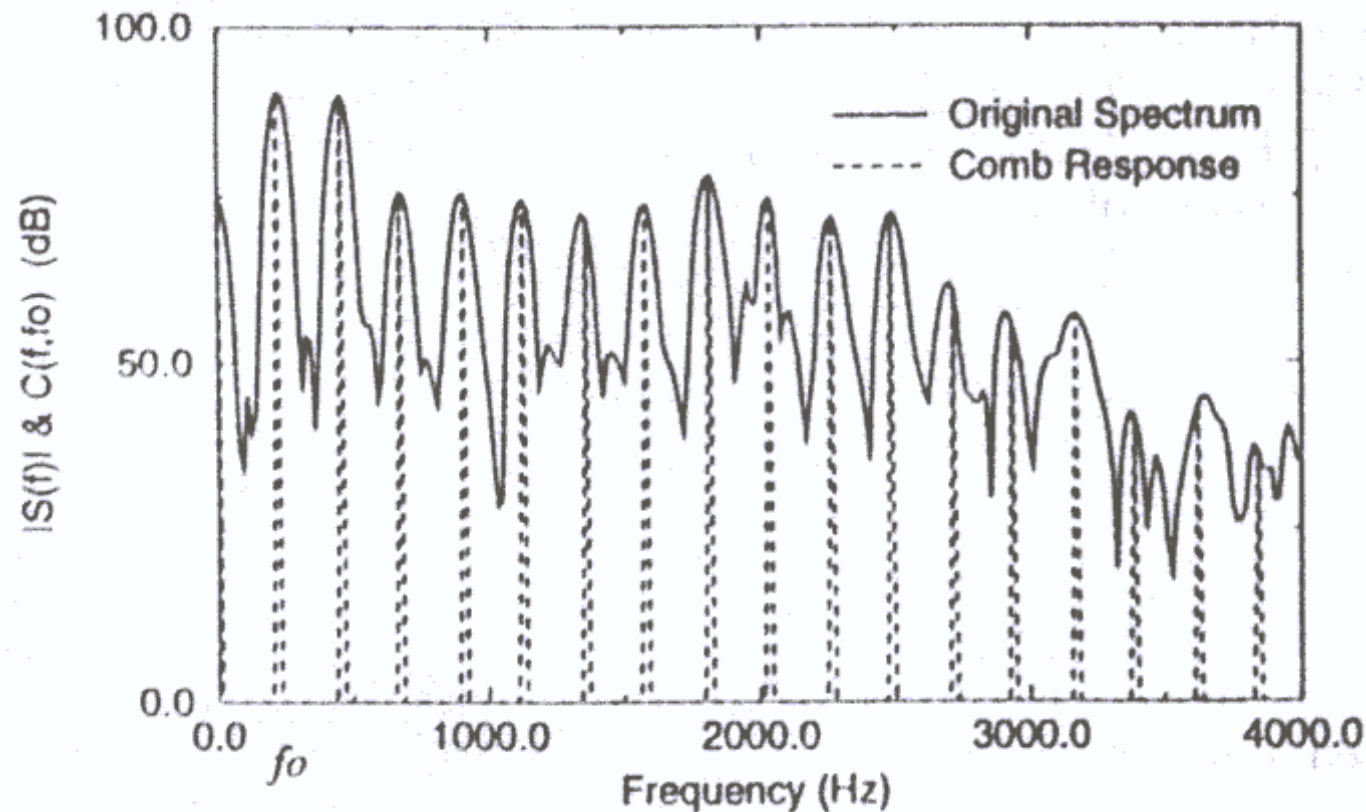
- ❑ Using comb filter in the frequency domain as in the following.
  - ❑ That is, to maximize the following autocorrelation output.

$$A_c(\omega_0) = \frac{\omega_0}{\Omega_m} \sum_{k=1}^{\Omega_m/\omega_0} S(k\omega_0)W(k\omega_0) \quad \frac{2\pi}{\tau_{\max}} \leq \omega_0 \leq \frac{2\pi}{\tau_{\min}}$$

- ❑ Here,  $\omega_0$ : fundamental freq.,  $\Omega_m$ :  $(2\pi f_s)/2$ ,  $W(k\omega_0)$ : comb peaks
- ❑ Actually, the first harmonic component is likely to disappear due to front-end filtering, therefore it is desirable to determine the period by utilizing the entire harmonics.

## ***Frequency domain methods for PD***

- ❑ Harmonic peak detection method using comb filter



# ***Frequency domain methods for PD***

## ❑ Spectrum similarity method

❑ Comparing the reconstructed spectrum with the original speech spectrum

❑ That is, to minimize the following error function.

$$E(\omega_0) = \sum_{m=0}^{M-1} \left( S(m) - \hat{S}(m, \omega_0) \right)^2$$

$$\hat{S}(m, \omega_0) = \begin{cases} A_0(\omega_0) W\left(\frac{2\pi}{M} m\right) \\ A_1(\omega_0) W\left(\frac{2\pi}{M} m - \omega_0\right) \\ \vdots \\ A_l(\omega_0) W\left(\frac{2\pi}{M} m - l\omega_0\right) \\ \vdots \end{cases}$$

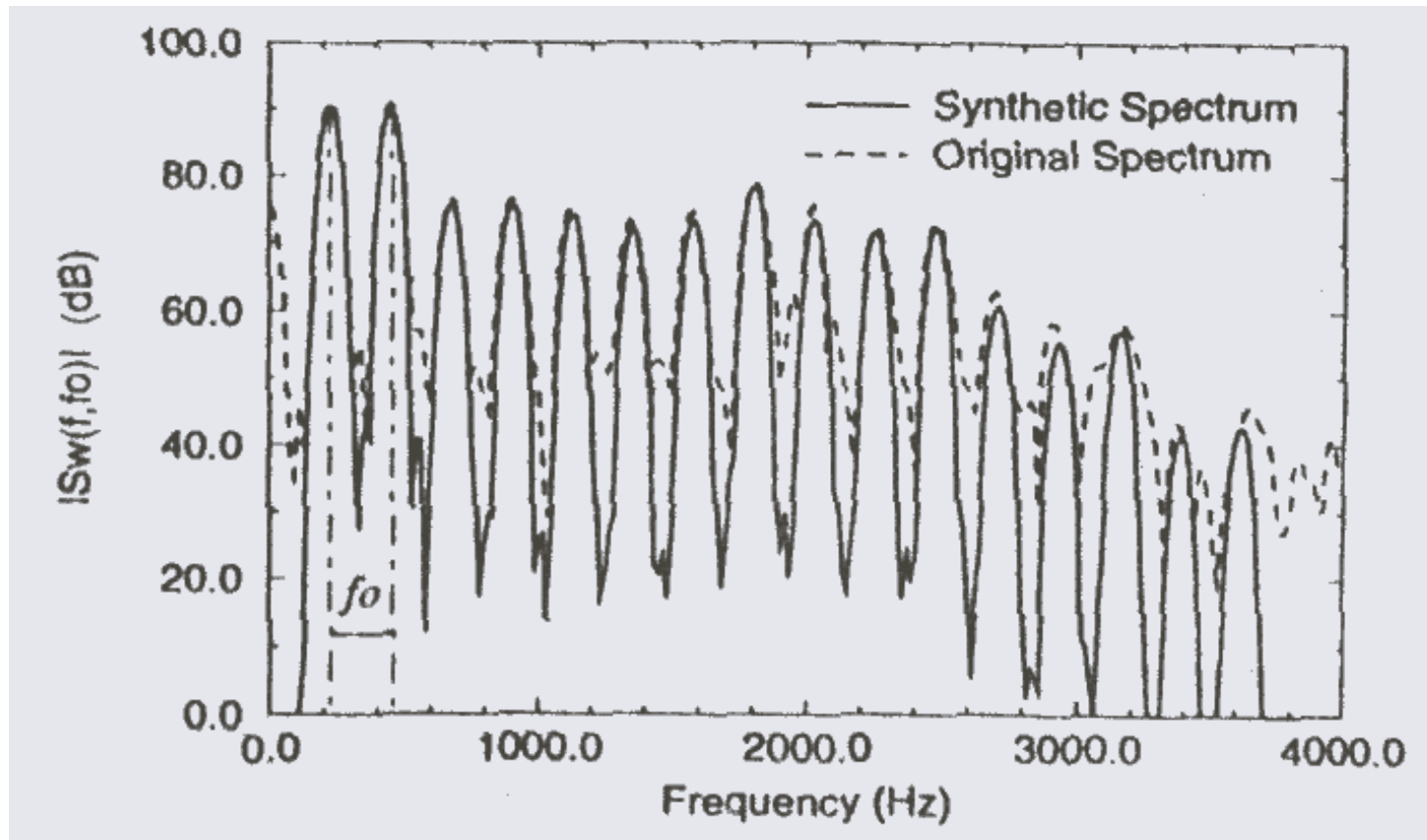
$$A_l(\omega_0) = \frac{\sum_{m=a_l}^{b_l} S(m) W\left(\frac{2\pi}{M} m - l\omega_0\right)}{\sum_{m=a_l}^{b_l} \left| W\left(\frac{2\pi}{M} m - l\omega_0\right) \right|^2}$$

$$a_l = \left\lceil \frac{M}{2\pi} \left( l - \frac{1}{2} \right) \omega_0 \right\rceil$$

$$b_l = \left\lfloor \frac{M}{2\pi} \left( l + \frac{1}{2} \right) \omega_0 \right\rfloor = a_{l+1} - 1$$

# ***Frequency domain methods for PD***

## ☐ Spectrum similarity method



# ***Time- and frequency-domain methods for PD***

## ❑ Pitch estimation using spectral autocorrelation

❑ Redefine the normalized autocorrelation function as follows.

$$R_T(\tau) = \frac{\sum_{n=0}^{N-\tau-1} s(n)s(n+\tau)}{\sqrt{\sum_{n=0}^{N-\tau-1} s^2(n) \sum_{n=0}^{N-\tau-1} s^2(n+\tau)}}$$

Called normalized temporal autocorrelation (TA)

❑ Similarly, define the normalized spectral autocorrelation (SA) function in the frequency domain.

$$S(m) = A(m)e^{j\theta(m)} \quad \text{for } 0 \leq m \leq M-1$$

$A_z(m)$ : zero-crossing spectrum

$$\omega_\tau = \lfloor M / \tau + 0.5 \rfloor$$

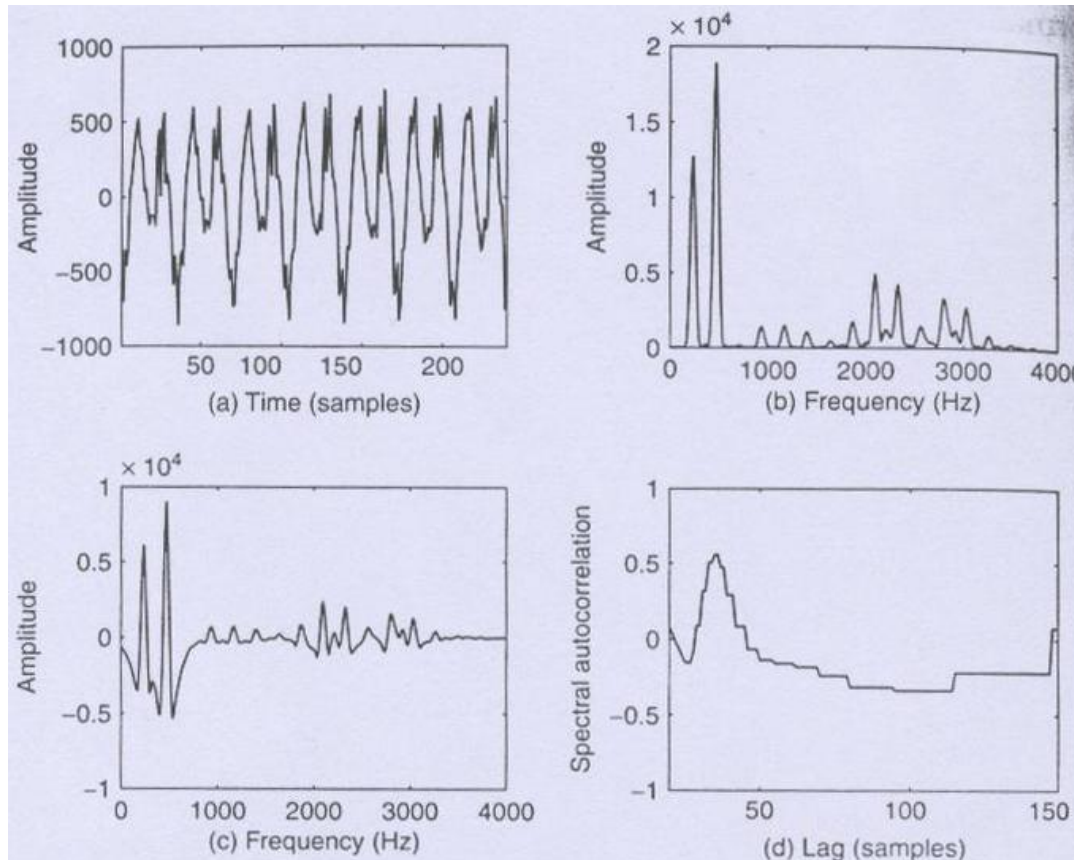
$T_0^{(l)}$  and  $T_0^{(u)}$ : lower and upper limits

$$R_S(\tau) = \frac{\sum_{m=0}^{\lfloor M/2 \rfloor - \omega_\tau} A_z(m)A_z(m + \omega_\tau)}{\sqrt{\sum_{m=0}^{\lfloor M/2 \rfloor - \omega_\tau} A_z^2(m) \sum_{m=0}^{\lfloor M/2 \rfloor - \omega_\tau} A_z^2(m + \omega_\tau)}} \quad \text{for } T_0^{(l)} \leq \tau \leq T_0^{(u)}$$



# ***Time- and frequency-domain methods for PD***

- ❑ Example of pitch estimation using spectral autocorrelation
  - ❑  $T_0 = 34$ -sample (as in female speech)

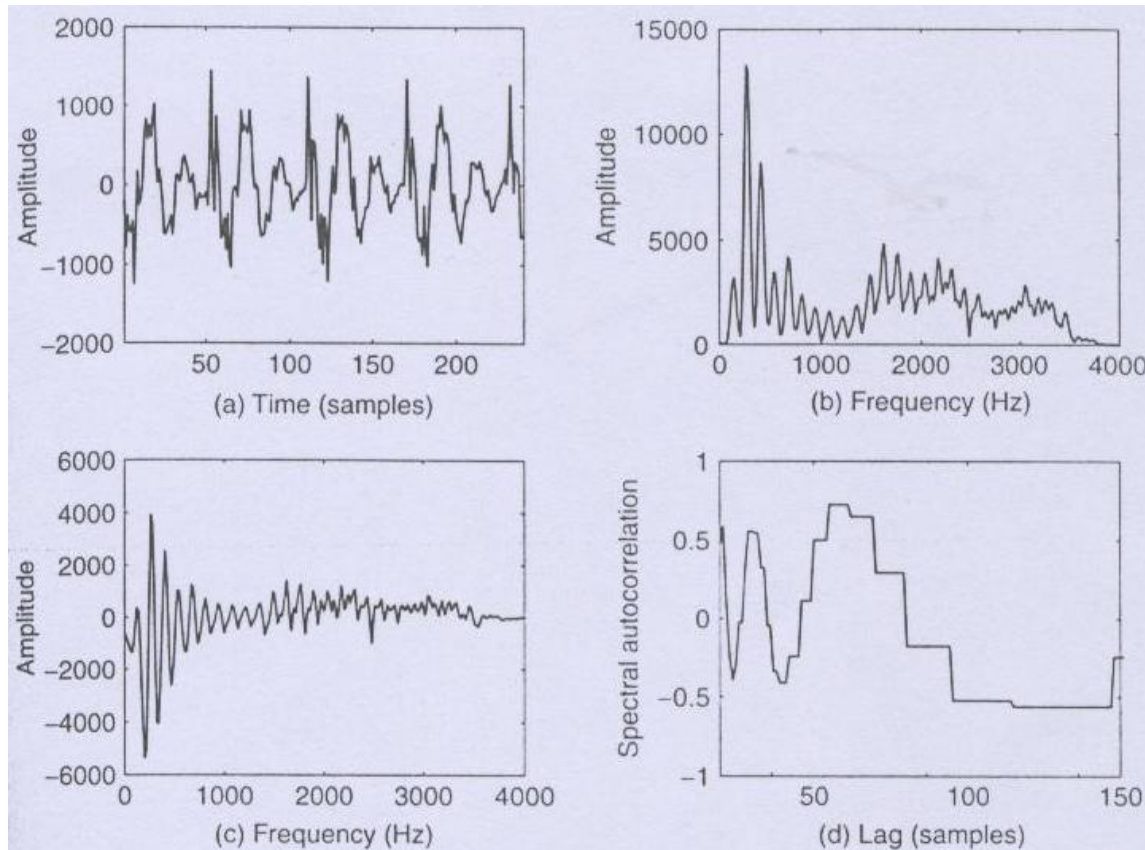


- (a) Speech signal (8 kHz)
- (b) Magnitude spectrum
- (c) Zero-crossing spectrum
- (d) SA function

→ Good result

# ***Time- and frequency-domain methods for PD***

- ❑ Example of pitch estimation using spectral autocorrelation
  - ❑  $T_0 = 59$ -sample (as in male speech)



- (a) Speech signal (8 kHz)
- (b) Magnitude spectrum
- (c) Zero-crossing spectrum
- (d) SA function

→ Not good result

## ***Time- and frequency-domain methods for PD***

- ❑ Analyzing the characteristics of the TA-based and SA-based PDAs, respectively,
  - ❑ TA-based PDA: likely to detect an unwanted pitch period multiple
  - ❑ SA-based PDA: likely to be pitch-halving
- ❑ Compensating for the problems by combining two methods,

$$R_{ST}(\tau) = \alpha R_T(\tau) + (1 - \alpha) R_S(\tau) \quad 0 \leq \alpha \leq 1$$

$$\hat{T}_0 = \arg \max_{\tau} \{R_{ST}(\tau)\}$$

- ❑ Called the spectro-temporal autocorrelation (STA) PDA

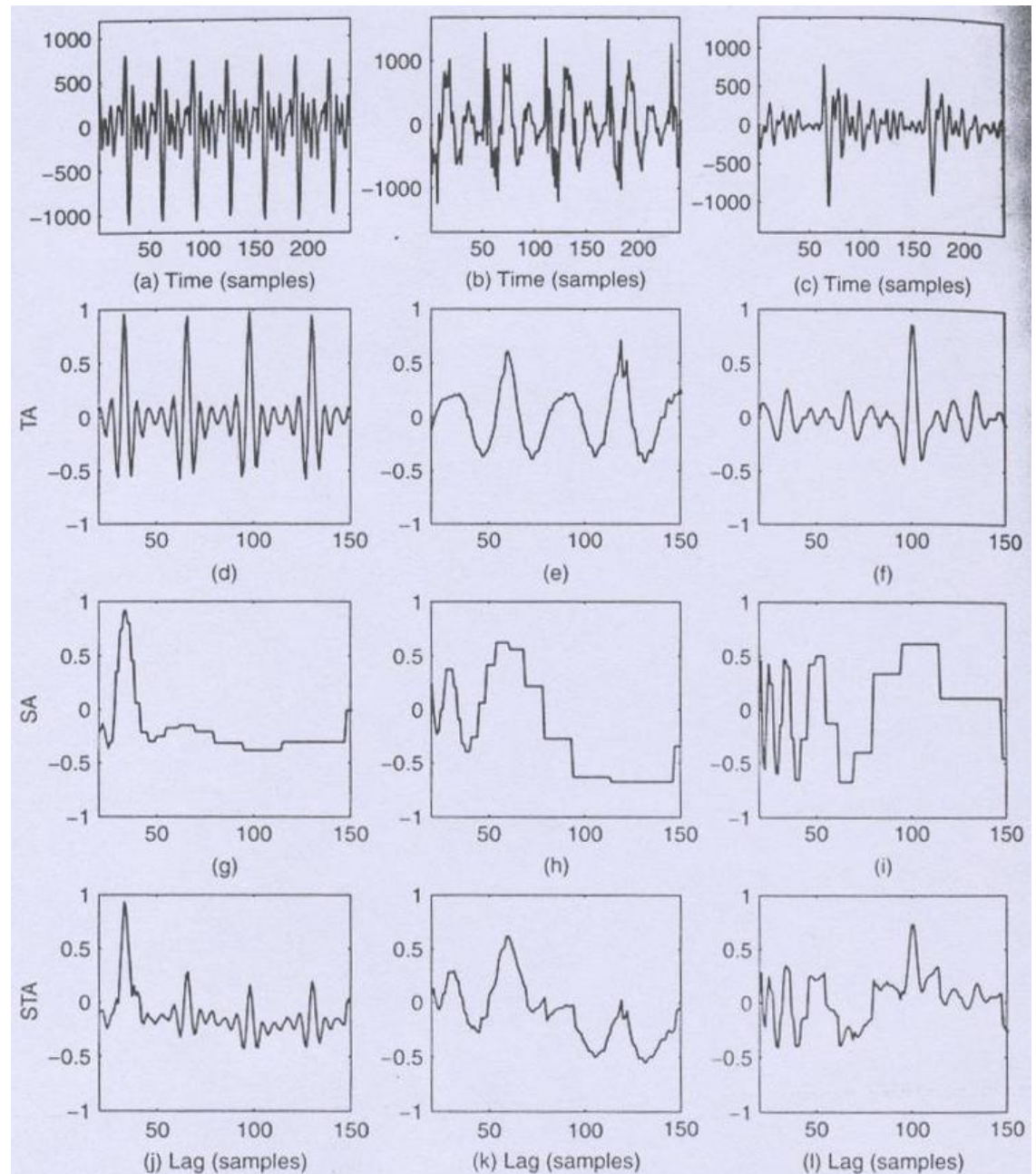
# ***Time- and frequency-domain methods for PD***

❑ Comparison of TA, SA, and STA ( $\alpha=0.5$ )

❑ Left: 32-sample  $T_0$

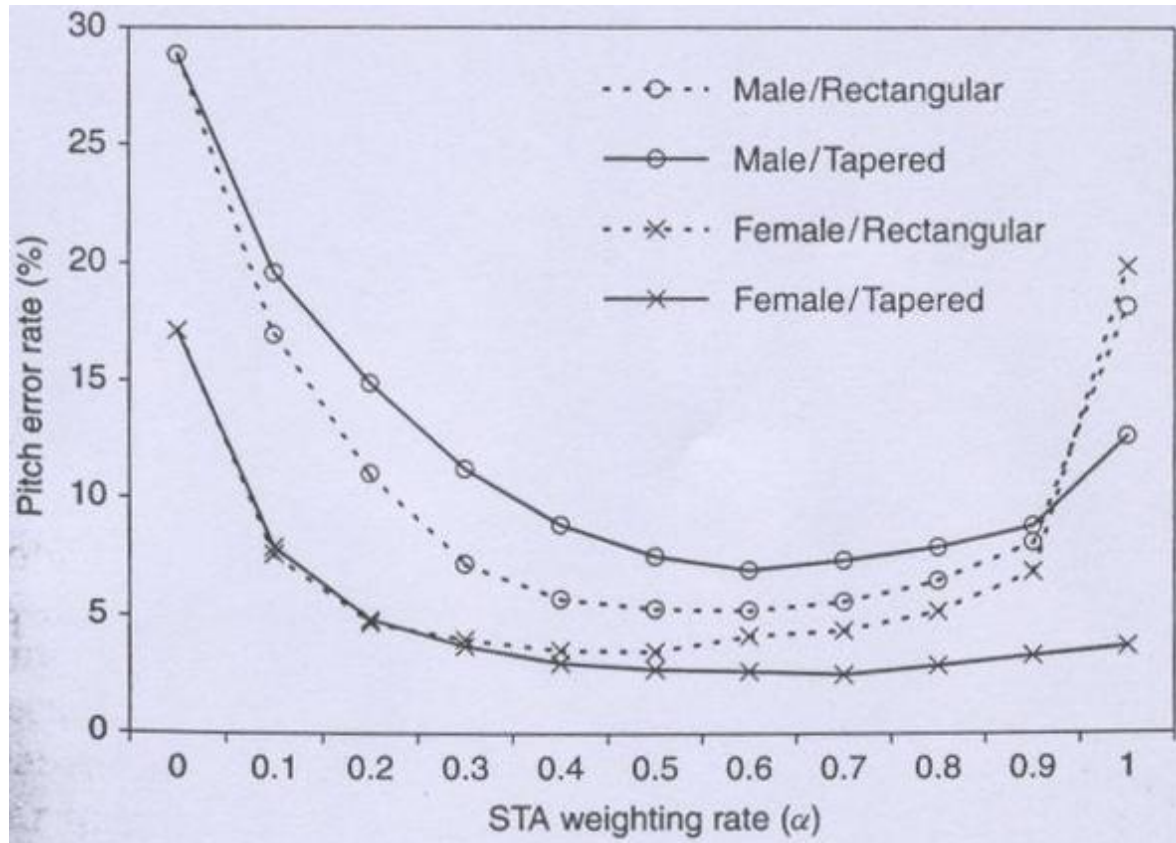
❑ Middle: 59-sample  $T_0$

❑ Right: 100-sample  $T_0$



## ***Time- and frequency-domain methods for PD***

- ❑ Analysis of the effect of the STA weighting factor  $\alpha$  in terms of the pitch error rate





# ***Pre- and post-processing techniques***

## **❑ Objectives**

- ❑ To improve the pitch period estimation performance

## **❑ Spectrum flattening**

- ❑ Removing the formants before pitch estimation process

- ❑ Linear method: using LPC inverse filter

- ❑ Drawback: The fundamental frequency and the first formant of high-pitch speech (like children or female) may be overlapped.  
→ This may destroy the entire periodicity information in the residual signals.

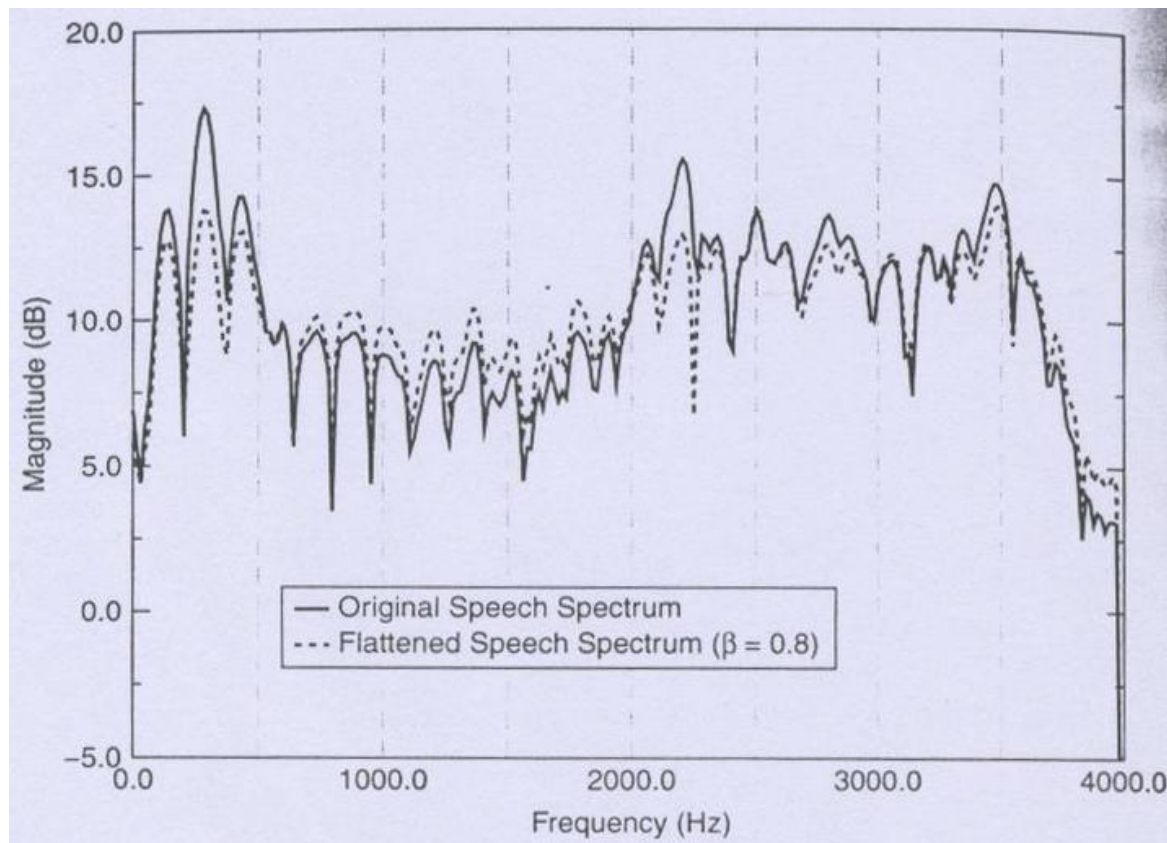
- ❑ Solution: obtaining the intermediate signal between the original and the LPC residual (even though high computations)

$$S_f(z) = \frac{A(z)}{A(z/\gamma)} S(z) \quad \text{for } 0 \leq \gamma \leq 1$$

- ❑  $S_f(z)$ : formant-suppressed signal,  $A(z)$ : inverse filter,  $\gamma$ : formant weighting factor

# ***Pre- and post-processing techniques***

- ❑ Spectrum flattening
  - ❑ Influence of the spectrum-flattening filter

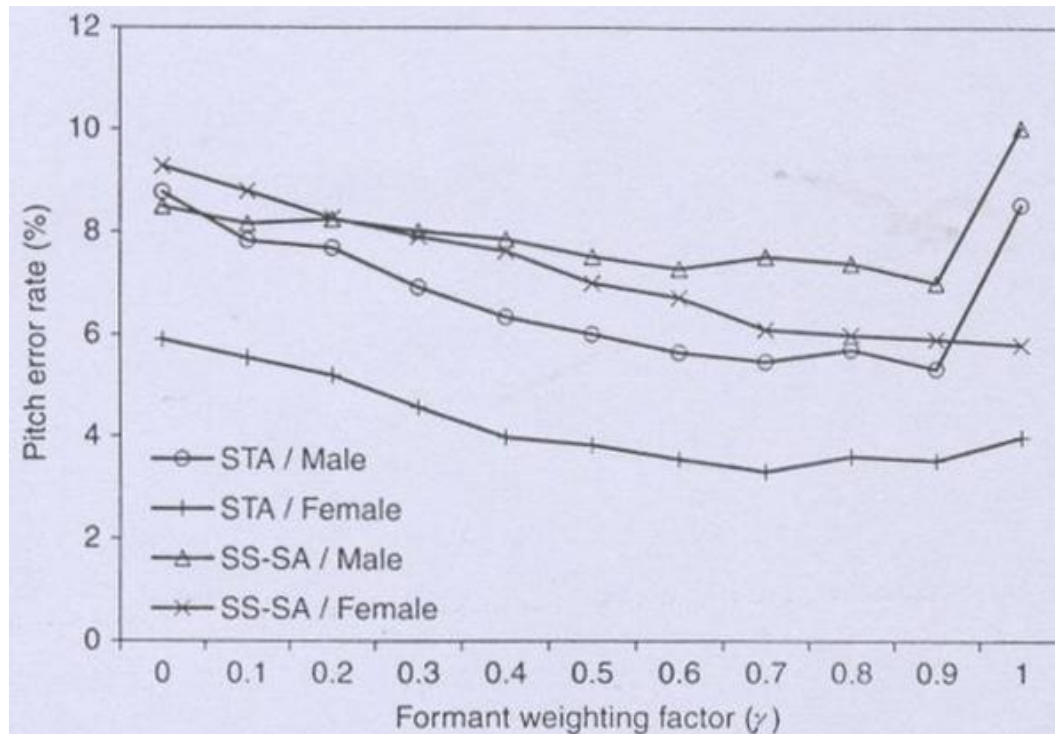


# ***Pre- and post-processing techniques***

## ❑ Spectrum flattening

❑ Analysis of the effect of  $\gamma$  in terms of the pitch error rate

❑ Here, SS-SA is a PDA using spectral synthesis – spectral autocorrelation method.



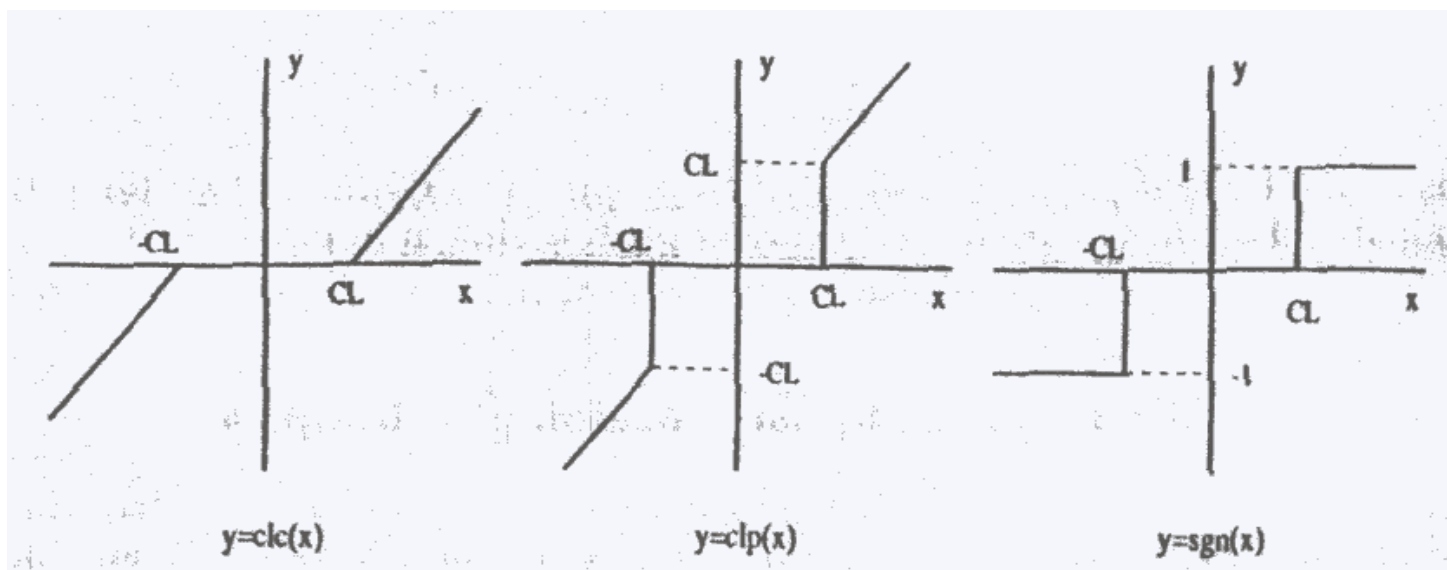


# ***Pre- and post-processing techniques***

## ❑ Spectrum flattening

❑ Non-linear method: using center clipping functions

❑ Several clipper functions for spectrum flattening



❑ Key problem: How to choose optimum clipping threshold (CL)

# ***Pre- and post-processing techniques***

## **❑ Pitch tracking**

- ❑ Principle: To utilize continuity characteristics of pitch in restricting the search space for pitch detection
  - ❑ For voiced speech, the variation of pitch period is small.
- ❑ Passive way: Smoothing the pitch periods after main determination
  - ❑ Drawback: Smoothing out an original abrupt change
- ❑ Active way: Applying a path penalty to main pitch determination process
  - ❑ Forward tracking & backward tracking
  - ❑ For example, once a pitch period of the current frame was estimated, the search for the pitch period of the next frame may be restricted to a range of a constant weighting of the current period.

# ***Pre- and post-processing techniques***

## **❑ Correction of multiple-pitch errors**

❑ Pitch determination process in time-domain PDA (e.g. auto-correlation method) probably results in those errors.

❑ First, a maximum peak is picked.

❑ Then, sub-multiple positions are checked by examining whether the ratio  $\frac{R(\tau_0 / i)}{R(\tau_0)} > TH$

❑ That is, if any, select a minimum integer  $i$  ( $\geq 2$ ) satisfying the above condition, and then determine  $\tau_0 / i$  as the final pitch period.

❑ There is no optimum solution. → The threshold is determined by tuning.

# ***Pre- and post-processing techniques***

## **❑ Correction of half-pitch errors**

- ❑ Pitch determination process in frequency-domain PDA (e.g. spectral auto-correlation method) probably results in those errors.
- ❑ Even in the time-domain PDA, if the previous ratio test is passed wrongly, pitch halving will take place.
- ❑ Therefore, for the vocoder sensitive to pitch period, another solution not using pitch detector is required.

# ***Voiced-unvoiced classification***

- ❑ Classifying the frame as either voiced or unvoiced
- ❑ Hard-decision voicing (binary voicing decision)
  - ❑ Periodic similarity (high for voiced)
  - ❑ Peakiness of speech (high)
  - ❑ Zero crossing rate (low)
  - ❑ Spectrum tilt (high)
  - ❑ Pre-emphasized energy ratio (low)
  - ❑ Low-band to full-band energy ratio (high)
  - ❑ Frame energy (high)
- ❑ Soft-decision voicing (mixed decision of voicing)
  - ❑ MBE mixed voicing
  - ❑ Split-band mixed voicing

# ***Hard-decision voicing***

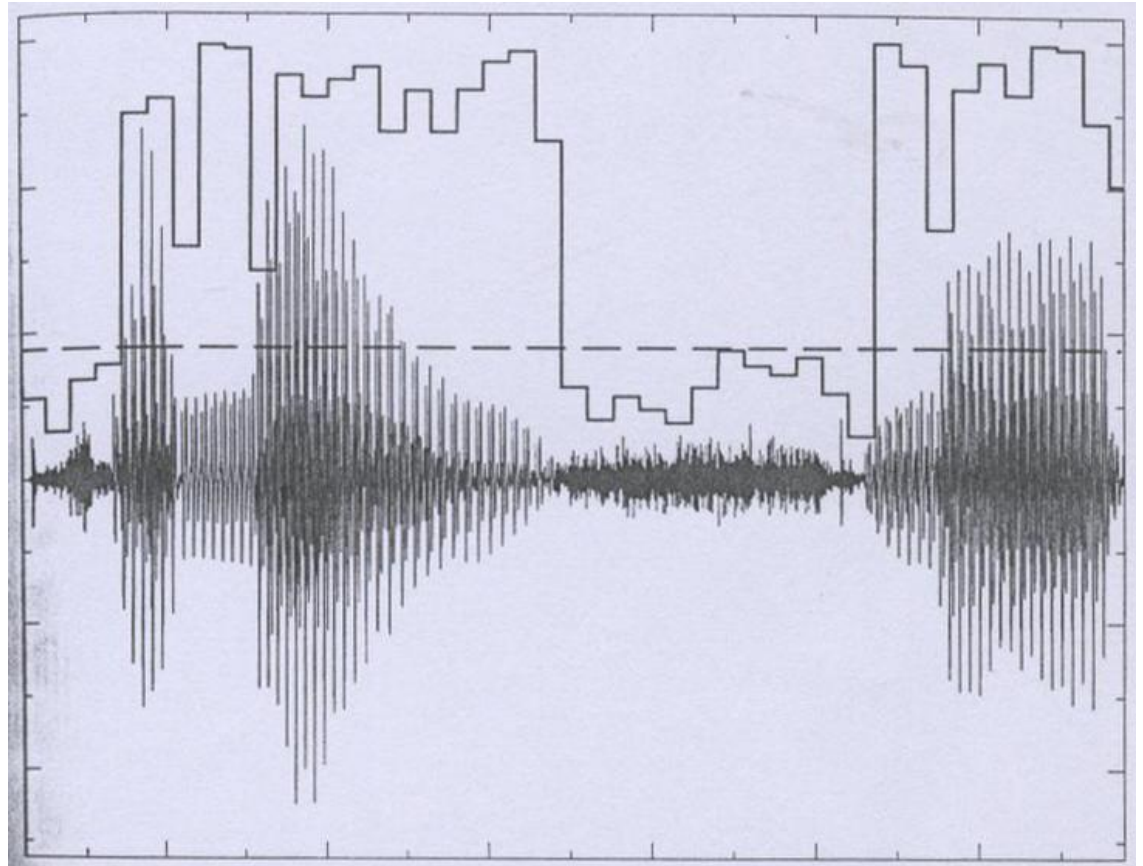
## ❑ Periodic similarity

❑ Measuring the regularity of waveform in terms of pitch period

$$Ps = \frac{\left[ \sum_{i=1}^N s(i)s(i-T) \right]^2}{\sum_{i=1}^N s^2(i) \sum_{i=1}^N s^2(i-T)}$$

❑  $T$ : pitch period

❑ A possible TH: 0.5



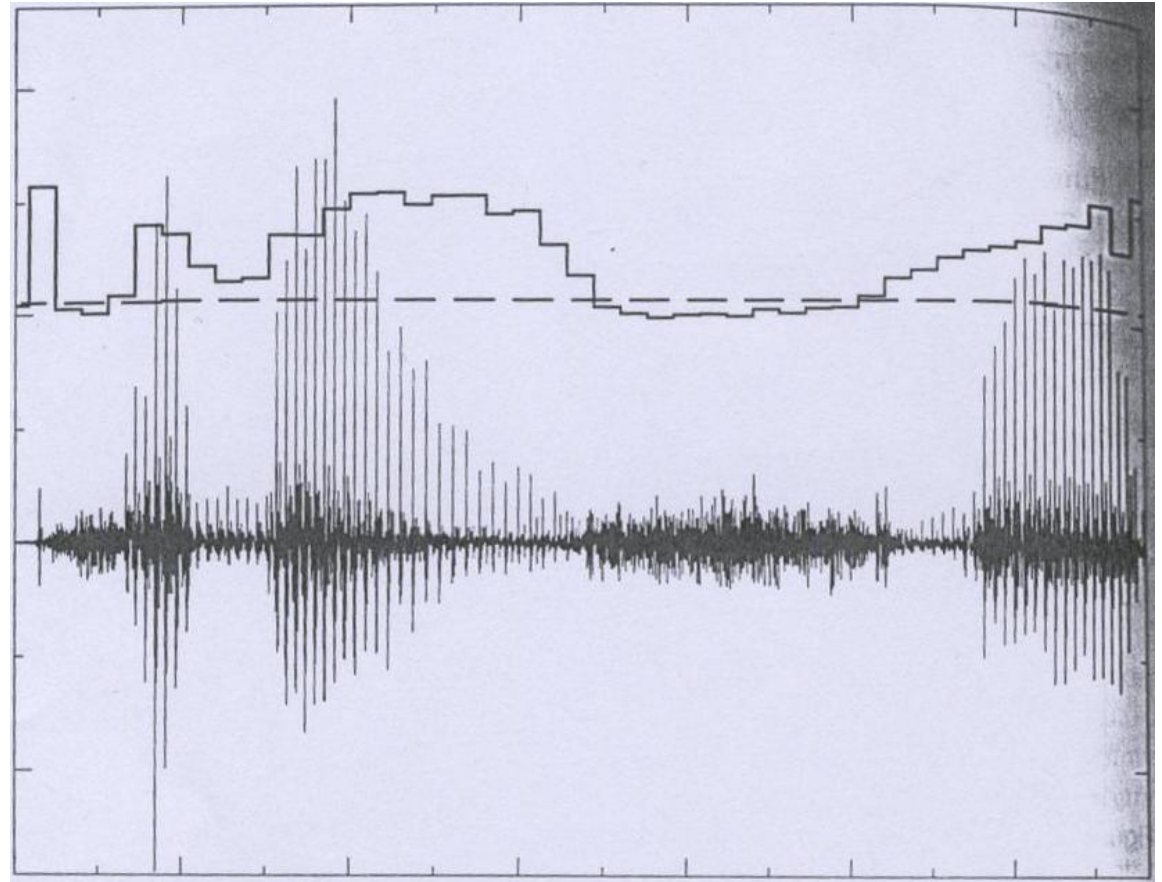
# ***Hard-decision voicing***

❑ Peakiness of speech

❑ Measuring the peakiness of the LPC residual

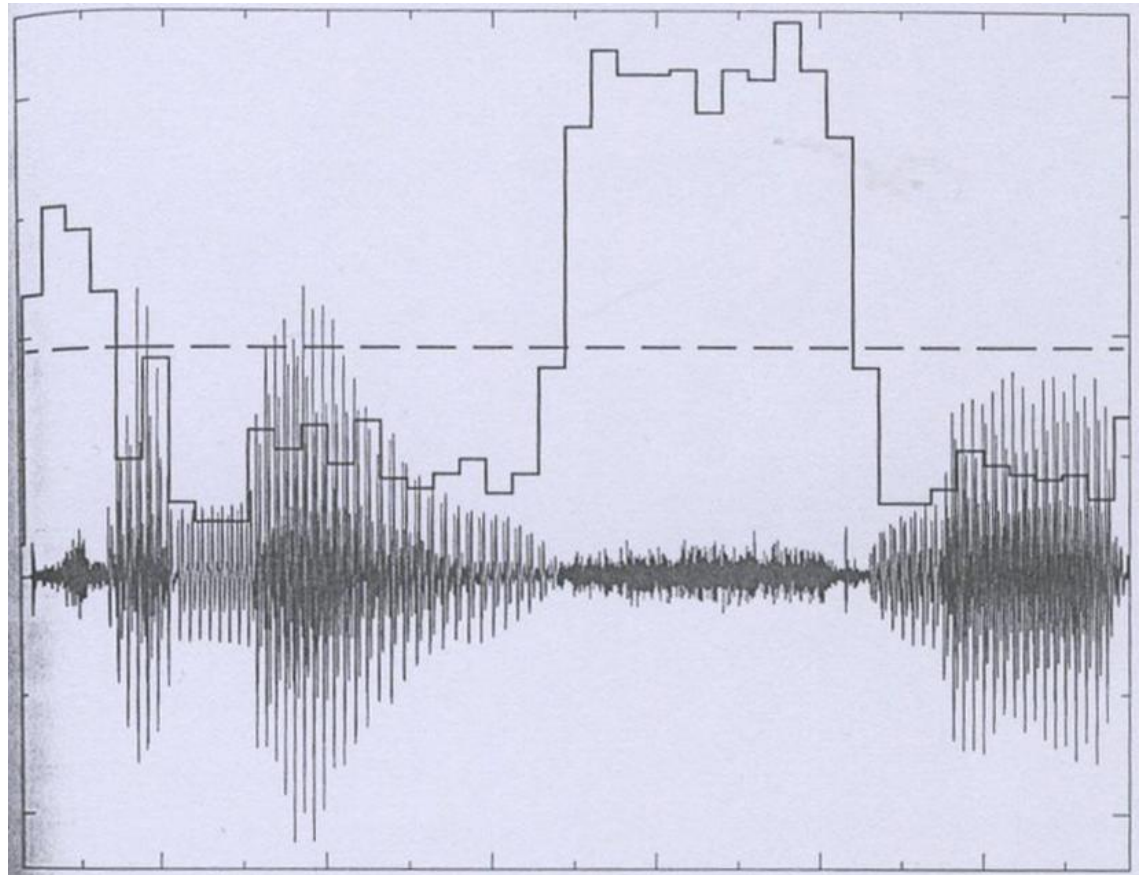
$$Pk = \frac{\sqrt{\frac{1}{N} \sum_{i=1}^N r^2(i)}}{\frac{1}{N} \sum_{i=1}^N |r(i)|}$$

❑ A possible TH: 1.4



# ***Hard-decision voicing***

- ❑ Zero crossing rate
  - ❑ Measuring the number of times the signal crosses the zero line
    - ❑ A possible TH: 60





# ***Hard-decision voicing***

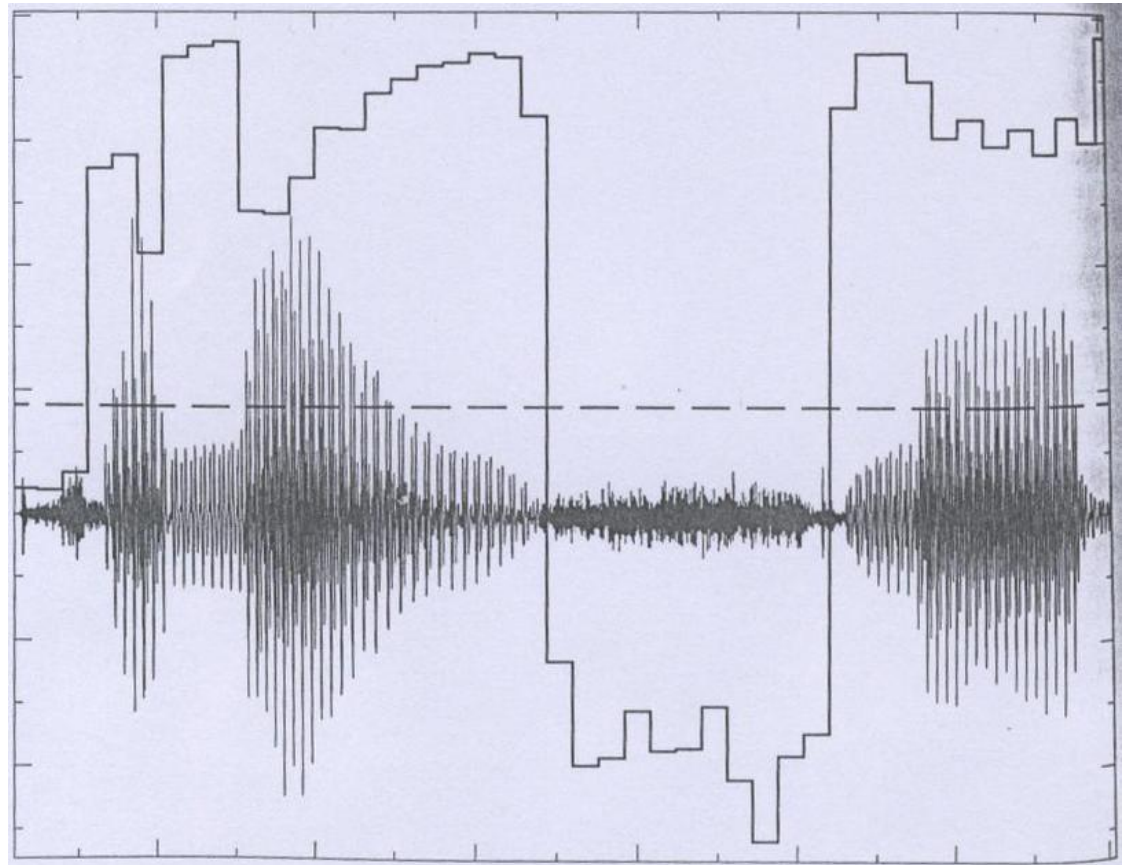
## ❑ Spectrum tilt

❑ Voiced speech has higher energy in low frequencies.

❑ Measuring the first-order normalized autocorrelation

$$St = \frac{\sum_{i=1}^N s(i)s(i-1)}{\sum_{i=1}^N s^2(i)}$$

❑ A possible TH: 0.25



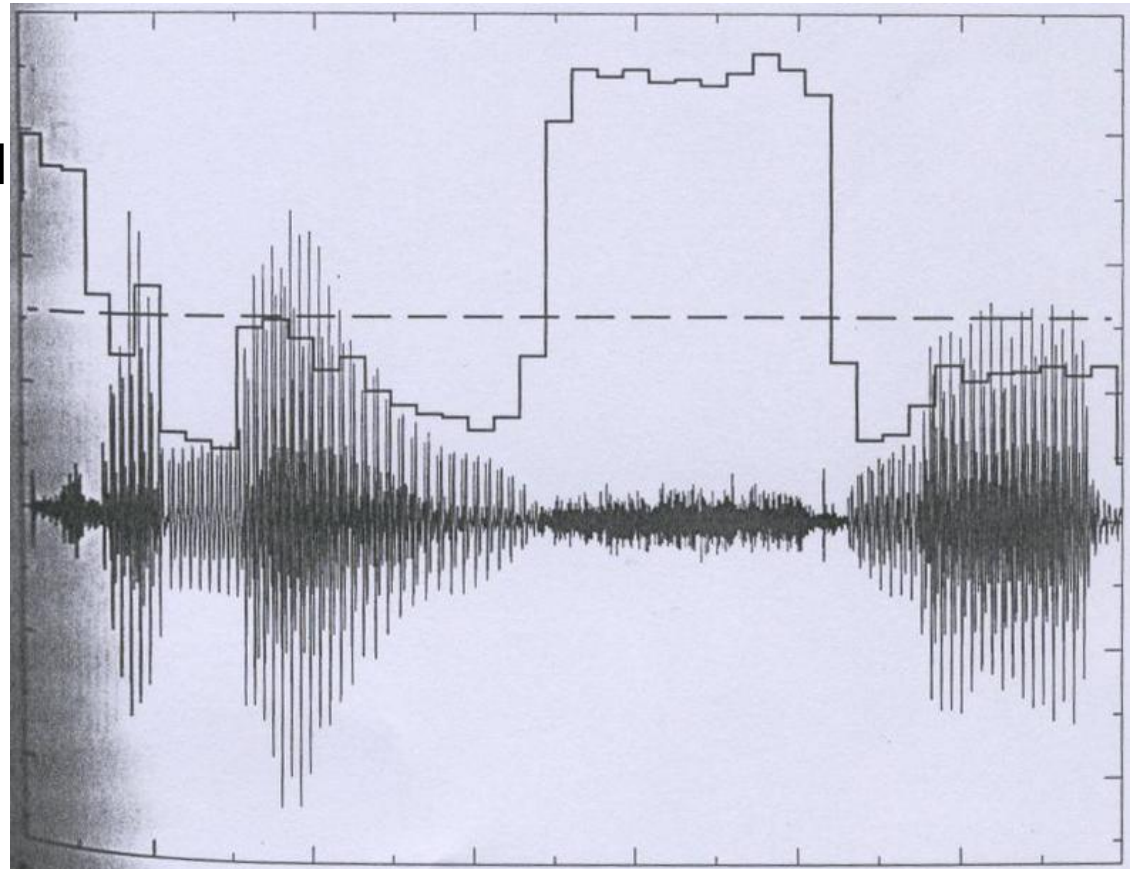
# ***Hard-decision voicing***

## ❑ Pre-emphasized energy ratio

- ❑ The first-order correlation of voiced samples is much higher than that of unvoiced.
- ❑ Measuring the ratio of the pre-emphasized energy to the original

$$Pr = \frac{\sum_{i=1}^N |s(i) - s(i-1)|}{\sum_{i=1}^N |s(i)|}$$

- ❑ A possible TH: 0.9

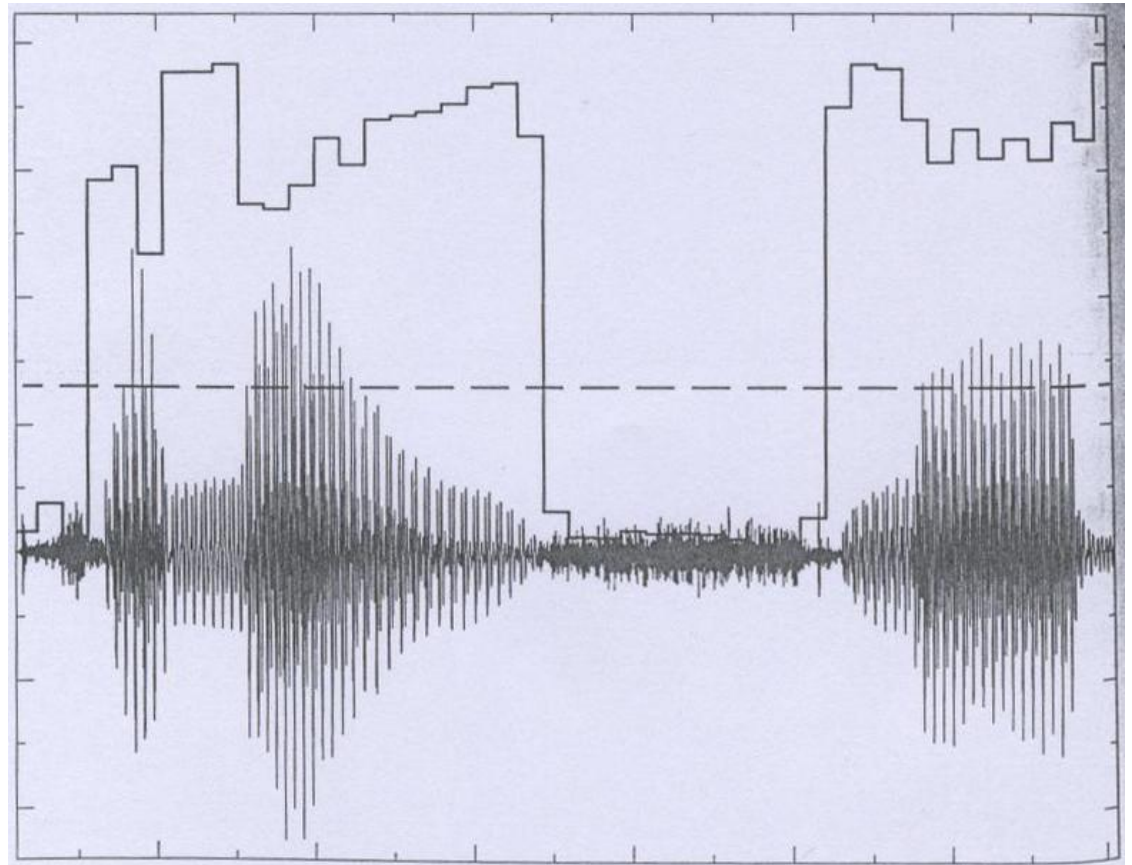


# ***Hard-decision voicing***

- ❑ Low-band to full-band energy ratio
  - ❑ Measuring the energy ratio of the first 1 kHz to the full-band energy

$$LF = \frac{\sum_{i=1}^N s_{lpf}^2(i)}{\sum_{i=1}^N s^2(i)}$$

- ❑ A possible TH: 0.4



# ***Hard-decision voicing***

## ❑ Frame energy

❑ Voiced speech usually has a higher energy not in the absolute value but in a relative amount.

❑ That is, a comparison of current frame energy with the tracked maximum and minimum energies, given as follow, would useful.

❑  $E_{\max}(n)$  can go up fast and come down slowly.

$$E_{\max}(n) = \begin{cases} \alpha E_{\max}(n-1) + (1-\alpha)E_0 & \text{if } E_0 > E_{\max}(n-1) \\ \gamma E_{\max}(n-1) + (1-\gamma)E_0 & \text{otherwise} \end{cases}$$

where  $E_0$  : current frame energy, and typically  $\alpha = 0.5$ ,  $\gamma = 0.98$

❑  $E_{\min}(n)$  can come down fast and go up slowly.

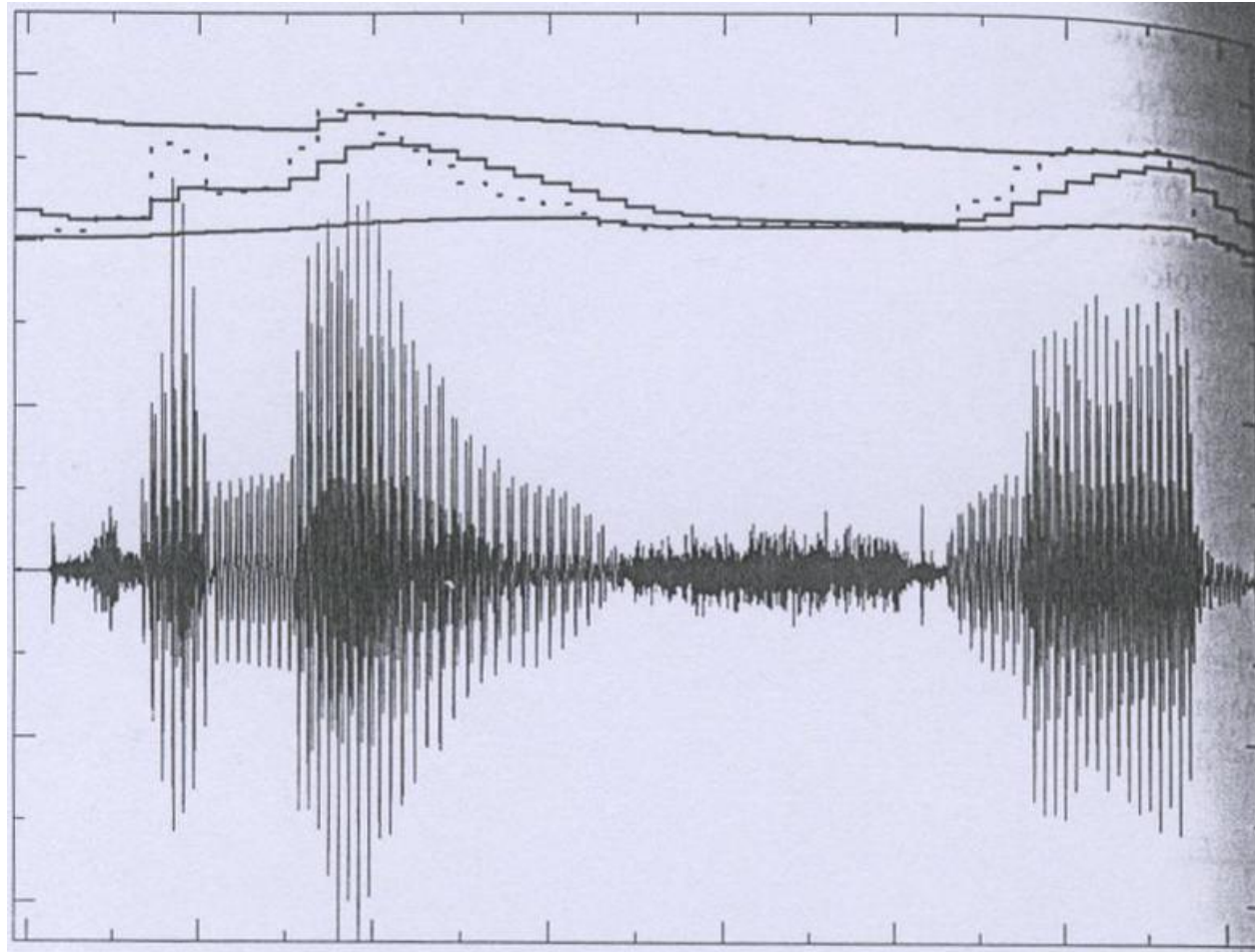
$$E_{\min}(n) = \begin{cases} \zeta E_{\min}(n-1) + (1-\zeta)E_0 & \text{if } E_0 < E_{\min}(n-1) \\ \beta E_{\min}(n-1) + (1-\beta)E_0 & \text{otherwise} \end{cases}$$

where typically  $\zeta = 0.55$ ,  $\beta = 0.98$

❑ Tracked average energy:  $E_{\text{av}}(n) = 0.75E_{\text{av}}(n-1) + 0.25E_0$

# ***Hard-decision voicing***

## ❑ Frame energy



$E_{\max}(n)$  track

$E_{\text{av}}(n)$  track

$E_{\min}(n)$  track

Dotted: frame energy

Speech waveform

Decision logic:

If  $\{(E_0 > E_{\max} - TH1) \text{ or } (E_0 > E_{\text{av}})\}$  Voiced,

Else if  $(E_0 < E_{\min} + TH2)$  Unvoiced,

Else Not-sure.

# ***Hard-decision voicing***

## ❑ Decision-making

### ❑ Combined decision using the voicing indicators

- ❑ The simplest way: majority vote
- ❑ Better rule: weighted combination

### ❑ Two-step normalization

- ❑ To compensate for differences of each parameter from the optimum decision threshold

$$P_s' = \begin{cases} (P_s - Th_{ps}) / (P_{s_{\max}} - Th_{ps}) & \text{if } P_s > Th_{ps} \\ (P_s - Th_{ps}) / (Th_{ps} - P_{s_{\min}}) & \text{if } P_s < Th_{ps} \end{cases}$$

$$Z_c' = \begin{cases} (Th_{zc} - Z_c) / (Th_{zc} - Z_{c_{\min}}) & \text{if } Z_c < Th_{zc} \\ (Th_{zc} - Z_c) / (Z_{c_{\max}} - Th_{zc}) & \text{if } Z_c > Th_{zc} \end{cases}$$

⋮

where  $Th_{ps}, Th_{pk}, Th_{zc}, \dots$  are fixed voicing thresholds



# ***Hard-decision voicing***

## ❑ Decision-making

### ❑ Two-step normalization (cont.)

- ❑ To compensate for different degrees of reliability, the overall voicing indicator  $V$  is

$$V = w_1Ps' + w_2Pk' + w_3Zc' + w_4St' + w_5LF' + w_6Pr' + w_7Fe'$$

- ❑ The weights are chosen according to the reliability of each indicator.

### ❑ Decision

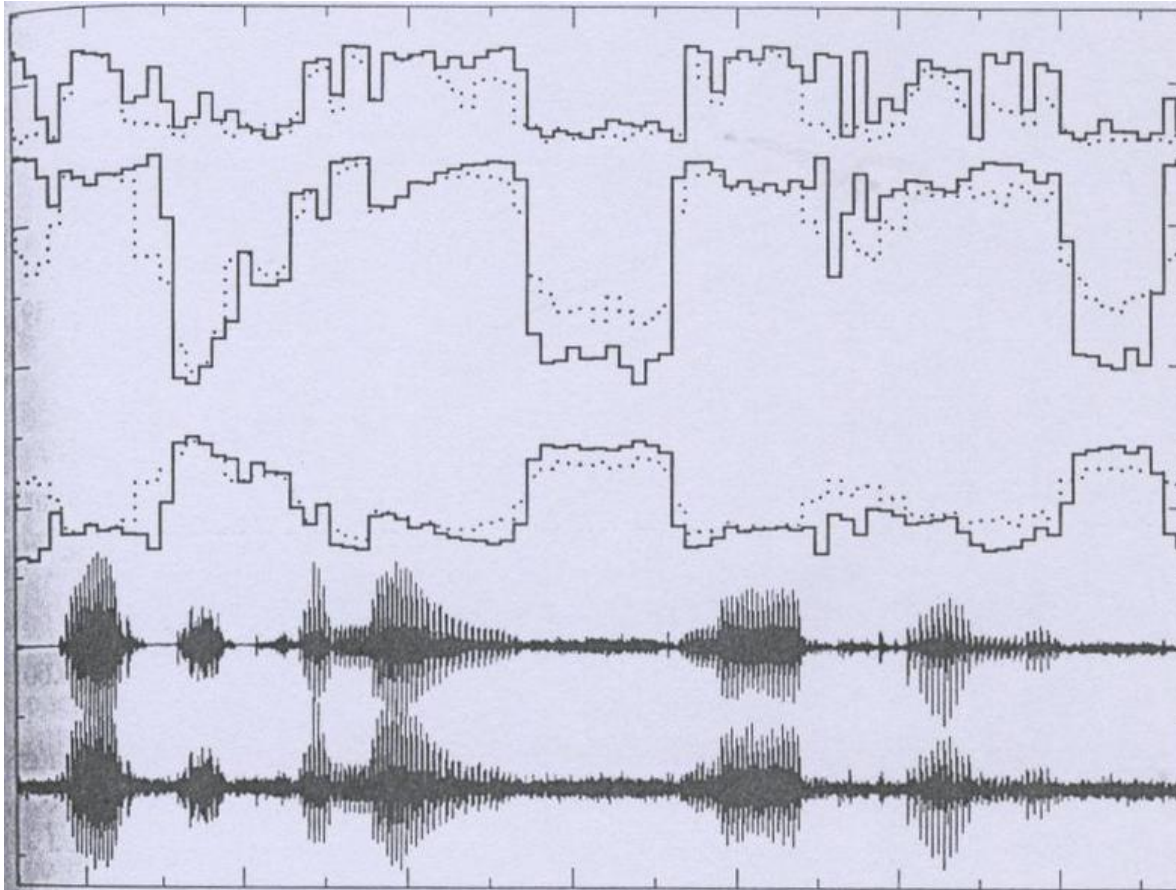
- ❑ When distinctively positive → voiced
- ❑ When distinctively negative → unvoiced
- ❑ If close to zero → unsure case → further checking

### ❑ Works very well with clean speech without background noises

# ***Hard-decision voicing***

## ❑ Problems

- ❑ When speech is mixed with background noise, the thresholds may not be valid anymore.
- ❑ When there is a transition from V to UV or vice versa even in clean speech,



$P_s$  plot

$P_r$  plot

$P_r$  plot

Clean speech  
waveform

Noisy speech with  
10 dB SNR vehicle  
noise

Dotted: the  
corresponding plots  
for noisy speech



## ***Soft-decision voicing***

- ❑ Alternative approach is to use a soft-decision voicing.
  - ❑ A frequency-domain voicing-decision process using the harmonic and random structures of voiced and unvoiced sounds, respectively
- ❑ Two methods
  - ❑ Multi-band excitation (MBE) mixed voicing
  - ❑ Split-band mixed voicing

# ***MBE mixed voicing***

## ❑ Voicing decision

- ❑ Define the normalized distance  $D_k$  between the original and the estimated speech spectra in each frequency band  $k$ .

$$D_k = \frac{\sum_{m=a_k}^{b_k} |S(m) - \hat{S}(m, \omega_0)|^2}{\sum_{m=a_k}^{b_k} |S(m)|^2}$$

- ❑  $\omega_0$ : the refined fundamental frequency after a post-processing
- ❑  $a_k, b_k$ : the first and last harmonic freq. bin indices in the  $k^{th}$  band
- ❑  $S(m)$ : the original speech spectrum
- ❑  $\hat{S}(m, \omega_0)$ : the reconstructed speech spectrum
- ❑ Bandwidth of each band: a multiple (e.g. 3) of  $\omega_0$ 
  - ❑ Thus, number of bands is dependent on the pitch period of the frame.

# ***MBE mixed voicing***

## ❑ Voicing decision

❑ The reconstructed speech spectrum is given by

$$\hat{S}(m, \omega_0) = \sum_{l=1}^L A_l(\omega_0) W_{l\omega_0}(m), \quad \lceil a_l \rceil \leq m < \lceil b_l \rceil$$

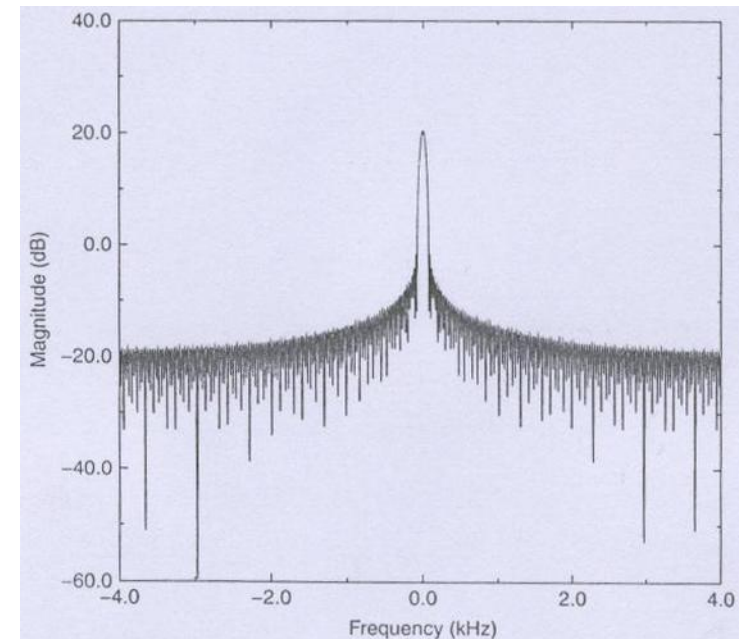
❑  $a_l = (l - 0.5)\omega_0$  ,  $b_l = (l + 0.5)\omega_0$

❑  $L$  : the number of harmonics within the 4 kHz bandwidth

❑  $W(m)$ : the frequency response of a suitable window that will be centered at the  $l^{th}$  harmonic of  $\omega_0$

❑  $A_l(\omega_0)$ : the  $l^{th}$  harmonic amplitude

$$A_l(\omega_0) = \frac{\sum_{m=\lceil a_l \rceil}^{\lceil b_l \rceil} S(m) W_{l\omega_0}(m)}{\sum_{m=\lceil a_l \rceil}^{\lceil b_l \rceil} |W_{l\omega_0}(m)|^2}$$



# ***MBE mixed voicing***

## □ Voicing decision

□ Compare with the adaptive threshold from listening tests

$$\Delta_k(\omega_0) = (\alpha + \beta\omega_0)[1.0 - \varepsilon(k-1)\omega_0]M(E_0, E_{\text{av}}, E_{\text{min}}, E_{\text{max}})$$

□  $\alpha = 0.35$ ,  $\beta = 0.557$ ,  $\varepsilon = 0.4775$  are the factors that give good subjective quality.

□  $M()$  is the adaptation factor that controls the decision threshold for V/UV decisions with  $\mu = 0.0075$ ,

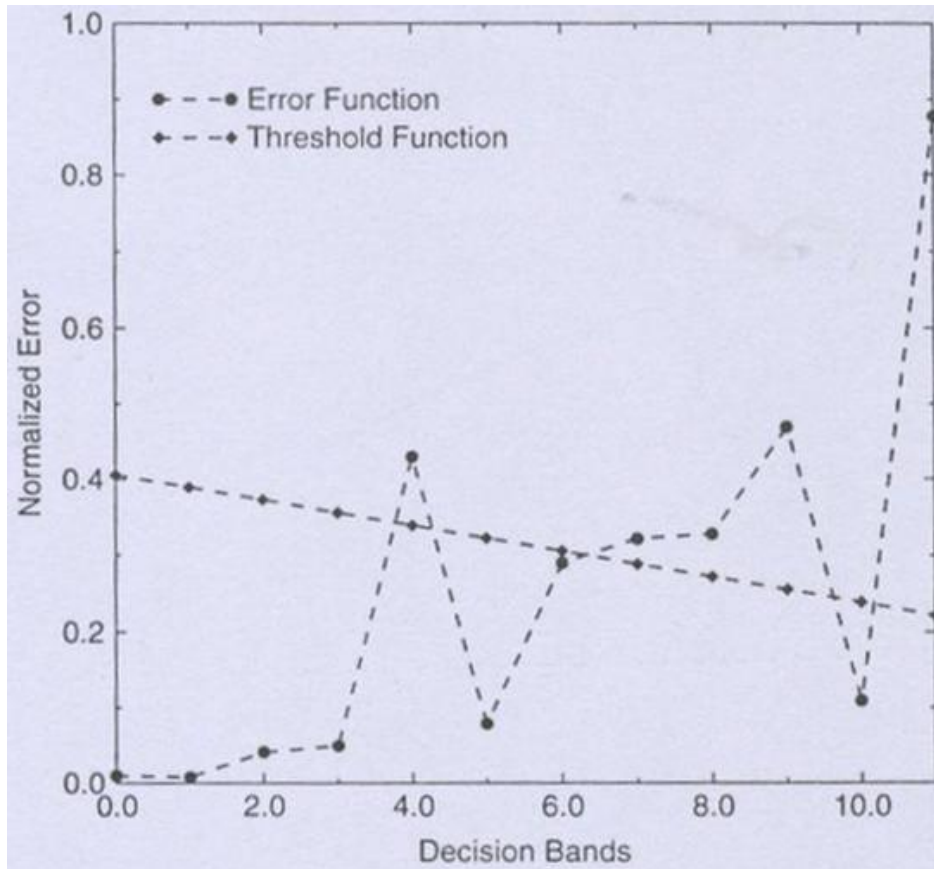
$$M(E_0, E_{\text{av}}, E_{\text{min}}, E_{\text{max}}) = \begin{cases} 0.5, & E_{\text{av}} < 200 \\ \frac{(E_0 + E_{\text{min}})(2E_0 + E_{\text{max}})}{(E_0 + \mu E_{\text{max}})(E_0 + E_{\text{max}})}, & E_{\text{av}} \geq 200 \text{ and } E_{\text{min}} < \mu E_{\text{max}} \\ 1.0, & \text{otherwise} \end{cases}$$

□ So, if  $D_k < \Delta_k(\omega_0)$ , then the band is regarded as voiced, elsewhere as unvoiced.

# ***MBE mixed voicing***

## ❑ Voicing decision

- ❑ Typical example of the error and threshold functions for one frame



From the threshold function, since  $\omega_0$  in male speech is relatively low, a lower band of male speech will be likely to be declared voiced, and a higher band of female speech will be likely to be declared unvoiced.

## ***Split-band mixed voicing***

- ❑ One drawback of MBE mixed voicing
  - ❑ More than one bit (12 bits in the previous) will be needed.
- ❑ Observation from experiments
  - ❑ If a spectrum contains an unvoiced band between two voiced bands, the unvoiced signal in the middle is usually small.
  - ❑ Thus if it is declared as voiced, subjectively it would not make much difference in speech quality.
- ❑ So, simply split the full band into low frequency band for voiced and high frequency band for unvoiced. → Split-band mixed voicing
  - ❑ Based on a more reliable measure such as voicing likelihood
  - ❑ Simply transmit the quantized voicing cut-off frequency.
    - ❑ Only 4 bits for the previous case

# ***Summary of lecture***

## ☐ Pitch estimation

- ☐ Detection of pitch period
- ☐ Time domain methods
  - ☐ AMDF, ACF, N-ACF
- ☐ Frequency domain methods
  - ☐ Harmonic peak detection method, Spectrum similarity method
- ☐ Time- and frequency-domain methods
  - ☐ Spectro-temporal autocorrelation (STA) PDA
- ☐ Pre- and post-processing techniques
  - ☐ Spectrum flattening, Pitch tracking, Correction of multiple- or half-pitch errors

## ☐ Voiced-unvoiced classification

- ☐ Hard-decision voicing
- ☐ Soft-decision voicing